

# Communication and Detection Theory

Signal and Information  
Processing Laboratory

Institut für Signal- und  
Informationsverarbeitung



Spring Semester 2017

Prof. Dr. A. Lapidoth

## Model Answers to Exercise 2 of February 28, 2017

<http://www.isi.ee.ethz.ch/teaching/courses/cdt>

---

### Problem 1

### *Reflection of Passband Signal*

- (i) Since  $\mathbf{x}_{\text{PB}}$  is the baseband representation of  $\mathbf{x}_{\text{PB}}$ ,

$$x_{\text{PB}}(t) = 2 \operatorname{Re}(x_{\text{BB}}(t) e^{i2\pi f_c t}), \quad t \in \mathbb{R}.$$

Consequently, we can express  $\tilde{\mathbf{x}}_{\text{PB}}$  in the more suggestive way

$$\begin{aligned} \tilde{x}_{\text{PB}}(t) &= x_{\text{PB}}(-t) \\ &= 2 \operatorname{Re}(x_{\text{BB}}(-t) e^{-i2\pi f_c t}) \\ &= 2 \operatorname{Re}(x_{\text{BB}}^*(-t) e^{i2\pi f_c t}) \\ &= 2 \operatorname{Re}(\tilde{x}_{\text{BB}}^*(t) e^{i2\pi f_c t}), \quad t \in \mathbb{R}, \end{aligned} \tag{1}$$

where in the third equality we used the fact that the real part of a complex number is equal to the real part of its complex conjugate. Since  $\mathbf{x}_{\text{BB}}$  is bandlimited to  $W/2$  Hz, so is  $\tilde{\mathbf{x}}_{\text{BB}}^*$ . We thus conclude from (1) using Proposition 7.6.9 that the baseband representation of  $\tilde{\mathbf{x}}_{\text{PB}}$  is  $\tilde{\mathbf{x}}_{\text{BB}}^*$ .

- (ii) From the first part we know that the baseband representation of  $\mathbf{y}_{\text{PB}}$  is  $\tilde{\mathbf{y}}_{\text{BB}}^*$ . Consequently, by Theorem 7.6.10,

$$\begin{aligned} \langle \mathbf{x}_{\text{PB}}, \tilde{\mathbf{y}}_{\text{PB}} \rangle &= 2 \operatorname{Re}(\langle \mathbf{x}_{\text{BB}}, \tilde{\mathbf{y}}_{\text{BB}}^* \rangle) \\ &= 2 \operatorname{Re} \left( \int_{-\infty}^{\infty} x_{\text{BB}}(t) y_{\text{BB}}(-t) dt \right) \\ &= 2 \operatorname{Re} \left( (\mathbf{x}_{\text{BB}} \star \mathbf{y}_{\text{BB}})(0) \right). \end{aligned}$$

## Problem 2

## Symmetries of the FT

- (i) The proof relies on the fact that conjugation and integration commute (Proposition 2.3.1).

If  $\mathbf{x}$  is a real-valued signal with Fourier Transform  $\hat{x}(\cdot)$ , then

$$\begin{aligned}\hat{x}(f) &= \int_{-\infty}^{\infty} x(t)e^{-i2\pi ft} dt \\ &= \int_{-\infty}^{\infty} x^*(t) \left(e^{i2\pi ft}\right)^* dt \\ &= \left(\int_{-\infty}^{\infty} x(t)e^{i2\pi ft} dt\right)^* \\ &= \hat{x}^*(-f), \quad f \in \mathbb{R},\end{aligned}$$

where the first equality is the definition of the Fourier Transform; the second follows from our assumption that  $\mathbf{x}$  is real and is thus equal to its complex conjugate; the third equality follows by swapping integration and conjugation; and the fourth equality follows again from the definition of the Fourier Transform.

- (ii) If  $\mathbf{x}$  is purely imaginary, then we can define the real signal  $\mathbf{y}$  as

$$y(t) = \frac{1}{i}x(t), \quad t \in \mathbb{R}.$$

Since  $\mathbf{y}$  is real, its FT  $\hat{y}$  is conjugate symmetric. And by the linearity of the FT,  $\hat{\mathbf{x}} = i\hat{\mathbf{y}}$ . Thus

$$\begin{aligned}\hat{x}(f) &= i\hat{y}(f) \\ &= (-i)^*\hat{y}(f) \\ &= -(i\hat{y}(-f))^* \\ &= -\hat{x}^*(-f), \quad f \in \mathbb{R}.\end{aligned}$$

- (iii) Let  $\mathbf{x}$  be a complex integrable signal. Then  $\mathbf{x}$  can be expressed as a sum

$$\mathbf{x} = \mathbf{x}_R + \mathbf{x}_I,$$

where  $\mathbf{x}_R$  is the integrable real signal

$$x_R(t) = \operatorname{Re}(x(t)), \quad t \in \mathbb{R},$$

and where  $\mathbf{x}_I$  is the integrable purely-imaginary signal

$$x_I(t) = i \operatorname{Im}(x(t)), \quad t \in \mathbb{R}.$$

By the linearity of the FT,

$$\hat{\mathbf{x}} = \hat{\mathbf{x}}_R + \hat{\mathbf{x}}_I,$$

thus expressing the FT of  $\mathbf{x}$  as a sum of the conjugate symmetric function  $\hat{\mathbf{x}}_R$  (because  $\mathbf{x}_R$  is real) and the conjugate antisymmetric function  $\hat{\mathbf{x}}_I$  (because  $\mathbf{x}_I$  is purely imaginary).

To prove uniqueness, suppose that

$$\hat{x}(f) = g_{cs}(f) + g_{cas}(f), \quad f \in \mathbb{R}, \tag{2}$$

where  $\mathbf{g}_{\text{cs}}$  is conjugate symmetric

$$g_{\text{cs}}(-f) = g_{\text{cs}}^*(f), \quad f \in \mathbb{R}, \quad (3)$$

and  $\mathbf{g}_{\text{cas}}$  is conjugate antisymmetric

$$g_{\text{cas}}(-f) = -g_{\text{cas}}^*(f), \quad f \in \mathbb{R}. \quad (4)$$

Then

$$\begin{aligned} \hat{x}^*(-f) &= g_{\text{cs}}^*(-f) + g_{\text{cas}}^*(-f) \\ &= g_{\text{cs}}(f) - g_{\text{cas}}(f), \quad f \in \mathbb{R}, \end{aligned} \quad (5)$$

where the second equality follows from (3) and (4).

Adding (2) and (5) we obtain the unique solution

$$g_{\text{cs}}(f) = \frac{\hat{x}(f) + \hat{x}^*(-f)}{2}, \quad f \in \mathbb{R}, \quad (6)$$

and substituting (5) from (2) yields the unique solution

$$g_{\text{cas}}(f) = \frac{\hat{x}(f) - \hat{x}^*(-f)}{2}, \quad f \in \mathbb{R}. \quad (7)$$

### Problem 3

### Phase Shift

(i) We first express  $z_{\text{PB}}(\cdot)$  in the more suggesting form

$$\begin{aligned} z_{\text{PB}}(t) &= x(t) \sin(2\pi f_c t + \phi) \\ &= 2 \operatorname{Re} \left( -\frac{i}{2} x(t) e^{i(2\pi f_c t + \phi)} \right) \\ &= 2 \operatorname{Re} \left( -\frac{i}{2} x(t) e^{i\phi} e^{i2\pi f_c t} \right), \quad t \in \mathbb{R}, \end{aligned}$$

where in the second equality we used the fact that  $\mathbf{x}$  is real. It now follows from Proposition 7.6.9 that the baseband representation  $\mathbf{z}_{\text{BB}}$  of  $\mathbf{z}_{\text{PB}}$  is

$$\mathbf{z}_{\text{BB}} = -\frac{i}{2} e^{i\phi} \mathbf{x}.$$

(ii) By the previous part,

$$\hat{z}_{\text{BB}}(f) = -\frac{i}{2} e^{i\phi} \hat{x}(f), \quad f \in \mathbb{R}.$$

Consequently, by Proposition 7.6.8,

$$\hat{z}_{\text{PB}}(f) = -\frac{i}{2} e^{i\phi} \hat{x}(f - f_c) + \frac{i}{2} e^{-i\phi} \hat{x}^*(-f - f_c), \quad f \in \mathbb{R}.$$

Since  $\mathbf{x}$  is real, its FT is conjugate symmetric, so we can also write this as

$$\hat{z}_{\text{PB}}(f) = -\frac{i}{2} e^{i\phi} \hat{x}(f - f_c) + \frac{i}{2} e^{-i\phi} \hat{x}(f + f_c), \quad f \in \mathbb{R}.$$

**Problem 4**

*Purely Real and Purely Imaginary  
Baseband Representations*

- (i) Let  $\mathbf{x}_{\text{PB}}$  be a real integrable passband signal that is bandlimited to  $W$  Hz around the carrier frequency  $f_c$ . The FT  $\hat{\mathbf{x}}_{\text{BB}}$  of its baseband representation  $\mathbf{x}_{\text{BB}}$  is related to its FT  $\hat{\mathbf{x}}_{\text{PB}}$  via the relation

$$\hat{x}_{\text{BB}}(f) = \hat{x}_{\text{PB}}(f + f_c) \mathbb{I}\left\{|f| \leq \frac{W}{2}\right\}, \quad f \in \mathbb{R} \quad (8)$$

(Proposition 7.6.5). Consequently, the condition

$$\hat{x}_{\text{PB}}(f_c - \delta) = \hat{x}_{\text{PB}}^*(f_c + \delta), \quad |\delta| \leq \frac{W}{2}$$

is equivalent to the condition

$$\hat{x}_{\text{BB}}(-\delta) = \hat{x}_{\text{BB}}^*(\delta), \quad |\delta| \leq \frac{W}{2}. \quad (9)$$

It remains to argue that this latter condition is equivalent to the condition that  $\mathbf{x}_{\text{BB}}$  is real.

One direction is obvious: if  $\mathbf{x}_{\text{BB}}$  is real, then its FT is conjugate symmetric (Exercise 6.1).

The other direction is almost as obvious: The signal  $\mathbf{x}_{\text{BB}}$  is an integrable signal that is bandlimited to  $W/2$  Hz (Theorem 7.6.5). Consequently, it is equal to the IFT of its FT  $\hat{\mathbf{x}}_{\text{BB}}$  (Proposition 6.4.10). Thus, if (9) holds then

$$\begin{aligned} x_{\text{BB}}(t) &= \int_{-\frac{W}{2}}^{\frac{W}{2}} \hat{x}_{\text{BB}}(f) e^{i2\pi ft} df \\ &= \left( \int_{-\frac{W}{2}}^{\frac{W}{2}} \hat{x}_{\text{BB}}^*(f) e^{-i2\pi ft} df \right)^* \\ &= \left( \int_{-\frac{W}{2}}^{\frac{W}{2}} \hat{x}_{\text{BB}}(-f) e^{-i2\pi ft} df \right)^* \\ &= \left( \int_{-\frac{W}{2}}^{\frac{W}{2}} \hat{x}_{\text{BB}}(\tilde{f}) e^{i2\pi \tilde{f}t} d\tilde{f} \right)^* \\ &= x_{\text{BB}}^*(t), \quad t \in \mathbb{R}. \end{aligned}$$

- (ii) By (8), the condition

$$\hat{x}_{\text{PB}}(f_c - \delta) = -\hat{x}_{\text{PB}}^*(f_c + \delta), \quad |\delta| \leq \frac{W}{2}$$

is equivalent to the condition

$$\hat{x}_{\text{BB}}(-\delta) = -\hat{x}_{\text{BB}}^*(\delta), \quad |\delta| \leq \frac{W}{2}. \quad (10)$$

It remains to argue that this latter condition is equivalent to the condition that  $\mathbf{x}_{\text{BB}}$  is purely imaginary.

One direction is obvious: if  $\mathbf{x}_{\text{BB}}$  is purely imaginary, then its FT is conjugate antisymmetric (Exercise 6.1).

The other direction is almost as obvious: The signal  $\mathbf{x}_{\text{BB}}$  is an integrable signal that is bandlimited to  $W/2$  Hz (Theorem 7.6.5). Consequently, it is equal to the IFT of its FT  $\hat{\mathbf{x}}_{\text{BB}}$

(Proposition 6.4.10). Thus, if (10) holds then

$$\begin{aligned}
x_{\text{BB}}(t) &= \int_{-\frac{W}{2}}^{\frac{W}{2}} \hat{x}_{\text{BB}}(f) e^{i2\pi ft} \, df \\
&= \left( \int_{-\frac{W}{2}}^{\frac{W}{2}} \hat{x}_{\text{BB}}^*(f) e^{-i2\pi ft} \, df \right)^* \\
&= - \left( \int_{-\frac{W}{2}}^{\frac{W}{2}} \hat{x}_{\text{BB}}(-f) e^{-i2\pi ft} \, df \right)^* \\
&= - \left( \int_{-\frac{W}{2}}^{\frac{W}{2}} \hat{x}_{\text{BB}}(\tilde{f}) e^{i2\pi \tilde{f}t} \, d\tilde{f} \right)^* \\
&= -x_{\text{BB}}^*(t), \quad t \in \mathbb{R}.
\end{aligned}$$

### Problem 5

### *Symmetry around the Carrier Frequency*

Let  $\mathbf{x}_{\text{PB}}$  be a real integrable passband signal that is bandlimited to  $W$  Hz around the carrier frequency  $f_c$ .

- (i) If  $\mathbf{x}_{\text{PB}}$  is equal to the signal  $t \mapsto w(t) \cos(2\pi f_c t)$  for some real integrable signal  $\mathbf{w}$  that is bandlimited to  $W/2$  Hz, then, by Proposition 7.6.9,  $\mathbf{x}_{\text{BB}}$  is equal to  $\mathbf{w}/2$ , because

$$\begin{aligned}
x_{\text{PB}}(t) &= w(t) \cos(2\pi f_c t) \\
&= 2 \operatorname{Re} \left( \frac{w(t)}{2} e^{i2\pi f_c t} \right),
\end{aligned}$$

where the second equality follows because  $\mathbf{w}$  is real. In this case  $\mathbf{x}_{\text{PB}}$  has a real baseband representation, namely  $\mathbf{w}/2$ , and by Exercise 7.1

$$\hat{x}_{\text{PB}}(f_c + \delta) = \hat{x}_{\text{PB}}^*(f_c - \delta), \quad |\delta| \leq \frac{W}{2}. \quad (11)$$

Conversely, if (11) holds, then by Exercise 7.1 the baseband representation of  $\mathbf{x}_{\text{PB}}$  is real, so

$$\begin{aligned}
x_{\text{PB}}(t) &= 2 \operatorname{Re} \left( x_{\text{BB}}(t) e^{i2\pi f_c t} \right) \\
&= 2x_{\text{BB}}(t) \cos(2\pi f_c t),
\end{aligned}$$

and  $\mathbf{x}_{\text{PB}}$  has the desired representation with  $\mathbf{w} = 2\mathbf{x}_{\text{BB}}$ .

- (ii) If  $\mathbf{x}_{\text{PB}}$  is equal to the signal  $t \mapsto w(t) \sin(2\pi f_c t)$  for some real integrable signal  $\mathbf{w}$  that is bandlimited to  $W/2$  Hz, then, by Proposition 7.6.9,  $\mathbf{x}_{\text{BB}}$  is equal to  $\mathbf{w}/(2i)$ , because

$$\begin{aligned}
x_{\text{PB}}(t) &= w(t) \sin(2\pi f_c t) \\
&= 2 \operatorname{Re} \left( \frac{w(t)}{2i} e^{i2\pi f_c t} \right),
\end{aligned}$$

where the second equality follows because  $\mathbf{w}$  is real. In this case  $\mathbf{x}_{\text{PB}}$  has a purely imaginary baseband representation, namely  $\mathbf{w}/(2i)$ , and by Exercise 7.1

$$\hat{x}_{\text{PB}}(f_c + \delta) = -\hat{x}_{\text{PB}}^*(f_c - \delta), \quad |\delta| \leq \frac{W}{2}. \quad (12)$$

Conversely, if (12) holds, then by Exercise 7.1 the baseband representation of  $\mathbf{x}_{\text{PB}}$  is purely imaginary, so

$$\begin{aligned}x_{\text{PB}}(t) &= 2 \operatorname{Re}\left(x_{\text{BB}}(t)e^{i2\pi f_c t}\right) \\ &= 2ix_{\text{BB}}(t) \sin(2\pi f_c t),\end{aligned}$$

and  $\mathbf{x}_{\text{PB}}$  has the desired representation with  $\mathbf{w} = 2i\mathbf{x}_{\text{BB}}$ .