

Communication and Detection Theory

Signal and Information
Processing Laboratory

Institut für Signal- und
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Model Answers to Exercise 3 of March 7, 2017

<http://www.isi.ee.ethz.ch/teaching/courses/cdt>

Problem 1

Separation between Signals

- (i) Since the norm is nonnegative, the condition $\|\mathbf{v} - \mathbf{u}_1\|_2 = \|\mathbf{v} - \mathbf{u}_2\|_2$ is equivalent to the condition

$$\|\mathbf{v} - \mathbf{u}_1\|_2^2 = \|\mathbf{v} - \mathbf{u}_2\|_2^2. \quad (1)$$

Moreover,

$$\|\mathbf{v} - \mathbf{u}_1\|_2^2 = \|\mathbf{v}\|_2^2 + \|\mathbf{u}_1\|_2^2 - 2 \operatorname{Re}(\langle \mathbf{v}, \mathbf{u}_1 \rangle), \quad (2)$$

$$\|\mathbf{v} - \mathbf{u}_2\|_2^2 = \|\mathbf{v}\|_2^2 + \|\mathbf{u}_2\|_2^2 - 2 \operatorname{Re}(\langle \mathbf{v}, \mathbf{u}_2 \rangle). \quad (3)$$

Taking the difference of (3) and (2) we see that the relation defining the subspace (1) can be written equivalently as

$$0 = \|\mathbf{u}_2\|_2^2 - \|\mathbf{u}_1\|_2^2 - 2 \operatorname{Re}(\langle \mathbf{v}, \mathbf{u}_2 - \mathbf{u}_1 \rangle)$$

or

$$\operatorname{Re}(\langle \mathbf{v}, \mathbf{u}_2 - \mathbf{u}_1 \rangle) = \frac{\|\mathbf{u}_2\|_2^2 - \|\mathbf{u}_1\|_2^2}{2}.$$

- (ii) In general, \mathcal{V} is *not* a linear subspace. Indeed, if $\|\mathbf{u}_1\|_2 \neq \|\mathbf{u}_2\|_2$ then the all-zero signal is not in \mathcal{V} .

- (iii) Let $\mathbf{w} = (\mathbf{u}_1 + \mathbf{u}_2)/2$. We have that

$$\|\mathbf{w} - \mathbf{u}_1\|_2 = \|(\mathbf{u}_2 - \mathbf{u}_1)/2\|_2,$$

$$\|\mathbf{w} - \mathbf{u}_2\|_2 = \|(\mathbf{u}_1 - \mathbf{u}_2)/2\|_2.$$

Since $\|(\mathbf{u}_2 - \mathbf{u}_1)/2\|_2 = \|(\mathbf{u}_1 - \mathbf{u}_2)/2\|_2$, we have that $\|\mathbf{w} - \mathbf{u}_1\|_2 = \|\mathbf{w} - \mathbf{u}_2\|_2$ and thus $\mathbf{w} = (\mathbf{u}_1 + \mathbf{u}_2)/2 \in \mathcal{V}$.

Problem 2

Orthogonal Subspace

Let \mathcal{U} denote the set of all signals $\mathbf{u} \in \mathcal{L}_2$ which are orthogonal to the signals $\mathbf{v}_1, \dots, \mathbf{v}_n$. By definition, the set \mathcal{U} is a subset of \mathcal{L}_2 . We have to show that it is closed under linear combinations.

For arbitrary \mathbf{u}_1 and \mathbf{u}_2 in \mathcal{U} and arbitrary complex numbers α_1 and α_2 , by the linearity of the inner product:

$$\langle \alpha_1 \mathbf{u}_1 + \alpha_2 \mathbf{u}_2, \mathbf{v}_\ell \rangle = \alpha_1 \langle \mathbf{u}_1, \mathbf{v}_\ell \rangle + \alpha_2 \langle \mathbf{u}_2, \mathbf{v}_\ell \rangle = 0, \quad \ell \in \{1, \dots, n\},$$

since

$$\langle \mathbf{u}_j, \mathbf{v}_\ell \rangle = 0, \quad j \in \{1, 2\}, \ell \in \{1, \dots, n\}.$$

Thus, any sum of the form $\alpha_1 \mathbf{u}_1 + \alpha_2 \mathbf{u}_2$ is an element of \mathcal{U} . This proves that \mathcal{U} is closed under linear combinations and hence forms a subspace of \mathcal{L}_2 .

Problem 3

Constructing an Orthonormal Basis

(i) See Figure 0.1

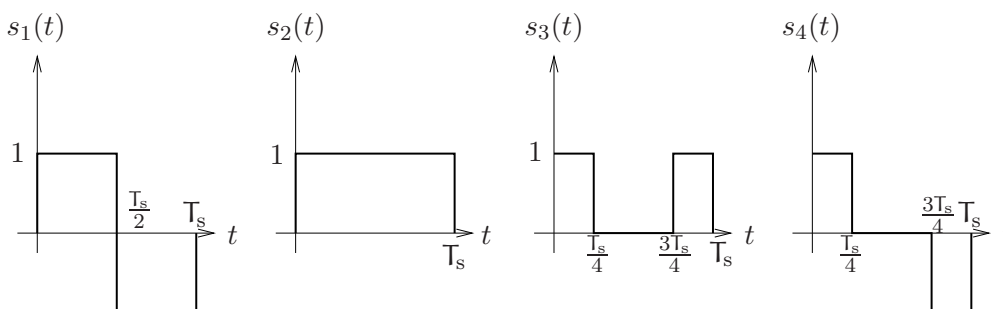


Figure 0.1:

(ii) It can be verified that the signals

$$\begin{aligned} \phi_1 &\triangleq \frac{\mathbf{s}_1}{\sqrt{T_s}}, \\ \phi_2 &\triangleq \frac{\mathbf{s}_2}{\sqrt{T_s}}, \\ \phi_3 &\triangleq \frac{2\mathbf{s}_3 - \mathbf{s}_2}{\sqrt{T_s}}, \\ \phi_4 &\triangleq \frac{2\mathbf{s}_4 - \mathbf{s}_1}{\sqrt{T_s}}, \end{aligned}$$

form an orthonormal basis for $\text{span}(\mathbf{s}_1, \mathbf{s}_2, \mathbf{s}_3, \mathbf{s}_4)$.

(iii) The signals $\mathbf{s}_1, \mathbf{s}_2, \mathbf{s}_3, \mathbf{s}_4$ can be expressed as:

$$\begin{aligned} \mathbf{s}_1 &= \sqrt{T_s} \phi_1, \\ \mathbf{s}_2 &= \sqrt{T_s} \phi_2, \\ \mathbf{s}_3 &= \frac{\sqrt{T_s}}{2} (\phi_3 + \phi_2), \\ \mathbf{s}_4 &= \frac{\sqrt{T_s}}{2} (\phi_4 + \phi_1). \end{aligned}$$

Problem 4

Constructing an Orthonormal Basis

Using (6.33) and (6.34) we obtain that

$$t \mapsto \text{sinc}\left(\frac{t}{2}\right) \text{ is of FT } f \mapsto 2\mathbf{I}\left\{|f| \leq \frac{1}{4}\right\}.$$

Consequently, by Proposition 6.6.1,

$$t \mapsto \text{sinc}^2\left(\frac{t}{2}\right) \text{ is of FT } f \mapsto (2 - 4|f|) \mathbb{I}\left\{|f| \leq \frac{1}{2}\right\},$$

i.e., of bandwidth 1/2 Hz. Consequently, if we define T as the reciprocal of twice the bandwidth, then $T = 1$. By Theorem 8.4.3 (i) we thus obtain the orthonormal expansion of $t \mapsto \text{sinc}^2(t/2)$

$$\sum_{\ell=-\infty}^{\infty} \text{sinc}^2\left(-\frac{\ell}{2}\right) \text{sinc}(t + \ell).$$

Since

$$\text{sinc}^2\left(-\frac{\ell}{2}\right) = \begin{cases} 1 & \text{if } \ell = 0, \\ 0 & \text{if } \ell \text{ is nonzero and even,} \\ 4/(\pi^2 \ell^2) & \text{if } \ell \text{ is odd,} \end{cases}$$

the expansion can also be written as

$$\text{sinc}(t) + \sum_{\ell \text{ odd}} \frac{4}{\pi^2 \ell^2} \text{sinc}(t + \ell).$$

Problem 5

Inner Product with a Bandlimited Signal

The signal \mathbf{y} can be written as

$$\begin{aligned} \mathbf{y} &= \mathbf{y} \star \text{LPF}_W + (\mathbf{y} - \mathbf{y} \star \text{LPF}_W) \\ &= \mathbf{y}_{\text{LPF}} + (\mathbf{y} - \mathbf{y}_{\text{LPF}}), \end{aligned}$$

where \mathbf{y}_{LPF} is defined as the convolution $\mathbf{y} \star \text{LPF}_W$ and, by Proposition 6.4.7, is an energy-limited signal that is bandlimited to W Hz. Consequently, by Theorem 8.7.1 Part (ii),

$$\langle \mathbf{x}, \mathbf{y}_{\text{LPF}} \rangle = T_s \sum_{\ell=-\infty}^{\infty} x(\ell T_s) y_{\text{LPF}}^*(\ell T_s).$$

Since the FT of $\mathbf{y} - \mathbf{y}_{\text{LPF}}$ vanishes in the band $[-W, W]$, and since \mathbf{x} is bandlimited to W Hz, it follows from Parseval's Theorem that

$$\langle \mathbf{x}, \mathbf{y} - \mathbf{y}_{\text{LPF}} \rangle = 0.$$

Hence

$$\begin{aligned} \langle \mathbf{x}, \mathbf{y} \rangle &= \langle \mathbf{x}, \mathbf{y}_{\text{LPF}} + \mathbf{y} - \mathbf{y}_{\text{LPF}} \rangle \\ &= \langle \mathbf{x}, \mathbf{y}_{\text{LPF}} \rangle + \langle \mathbf{x}, \mathbf{y} - \mathbf{y}_{\text{LPF}} \rangle \\ &= \langle \mathbf{x}, \mathbf{y}_{\text{LPF}} \rangle \\ &= T_s \sum_{\ell=-\infty}^{\infty} x(\ell T_s) y_{\text{LPF}}^*(\ell T_s). \end{aligned}$$

An alternative solution goes as follows. Let $T_s = 1/(2W)$. From the \mathcal{L}_2 -Sampling Theorem

$$\lim_{L \rightarrow \infty} \int_{-\infty}^{\infty} \left| x(t) - \sum_{\ell=-L}^L x(-\ell T_s) \text{sinc}\left(\frac{t}{T_s} + \ell\right) \right|^2 dt = 0.$$

For every $L \in \mathbb{N}$ define the signal

$$u_L(t) = x(t) - \sum_{\ell=-L}^L x(\ell T_s) \operatorname{sinc}\left(\frac{t}{T_s} + \ell\right), \quad t \in \mathbb{R}.$$

Then $\|\mathbf{u}_L\|_2 \rightarrow 0$ as $L \rightarrow \infty$. Fix some $\beta \in \mathbb{C}$. For every energy-limited \mathbf{y} we have by (4.14)

$$(\|\mathbf{y}\|_2 - |\beta| \|\mathbf{u}_L\|_2)^2 \leq \|\beta \mathbf{u}_L + \mathbf{y}\|_2^2 \leq (\|\mathbf{y}\|_2 + |\beta| \|\mathbf{u}_L\|_2)^2,$$

so $\|\beta \mathbf{u}_L + \mathbf{y}\|_2^2 \rightarrow \|\mathbf{y}\|_2^2$ as $L \rightarrow \infty$. We now expand $\|\beta \mathbf{u}_L + \mathbf{y}\|_2^2$:

$$\begin{aligned} \|\beta \mathbf{u}_L + \mathbf{y}\|_2^2 &= \int_{-\infty}^{\infty} \left| \beta \left(x(t) - \sum_{\ell=-L}^L x(-\ell T_s) \operatorname{sinc}\left(\frac{t}{T_s} + \ell\right) \right) + y(t) \right|^2 dt \\ &= |\beta|^2 \int_{-\infty}^{\infty} \left| x(t) - \sum_{\ell=-L}^L x(-\ell T_s) \operatorname{sinc}\left(\frac{t}{T_s} + \ell\right) \right|^2 dt + \int_{-\infty}^{\infty} |y(t)|^2 dt \\ &\quad + 2 \operatorname{Re} \left(\beta \int_{-\infty}^{\infty} x(t) y^*(t) dt - \beta \int_{-\infty}^{\infty} \sum_{\ell=-L}^L x(-\ell T_s) \operatorname{sinc}\left(\frac{t}{T_s} + \ell\right) y^*(t) dt \right) \\ &= |\beta|^2 \|\mathbf{u}_L\|_2^2 + \|\mathbf{y}\|_2^2 \\ &\quad + 2 \operatorname{Re} \left(\beta \langle \mathbf{x}, \mathbf{y} \rangle - \beta \sum_{\ell=-L}^L x(-\ell T_s) \int_{-\infty}^{\infty} \operatorname{sinc}\left(\frac{t}{T_s} + \ell\right) y^*(t) dt \right). \end{aligned}$$

Letting L tend to infinity and recalling that $\|\mathbf{u}_L\|_2 \rightarrow 0$ and $\|\beta \mathbf{u}_L + \mathbf{y}\|_2^2 \rightarrow \|\mathbf{y}\|_2^2$ we conclude that

$$2 \operatorname{Re} \left(\beta \langle \mathbf{x}, \mathbf{y} \rangle - \beta \sum_{\ell=-L}^L x(-\ell T_s) \int_{-\infty}^{\infty} \operatorname{sinc}\left(\frac{t}{T_s} + \ell\right) y^*(t) dt \right) = 0.$$

Since this has to hold for every $\beta \in \mathbb{C}$,

$$\langle \mathbf{x}, \mathbf{y} \rangle = \sum_{\ell=-L}^L x(-\ell T_s) \int_{-\infty}^{\infty} \operatorname{sinc}\left(\frac{t}{T_s} + \ell\right) y^*(t) dt. \quad (4)$$

We further have

$$\begin{aligned} \int_{-\infty}^{\infty} \operatorname{sinc}\left(\frac{t}{T_s} + \ell\right) y^*(t) dt &= \int_{-\infty}^{\infty} \operatorname{sinc}\left(\frac{1}{T_s}(-\ell T_s - t)\right) y^*(t) dt \\ &= \left(\mathbf{y}^* \star \left(t \mapsto \operatorname{sinc}\left(\frac{t}{T_s}\right) \right) \right)(-\ell T_s) \\ &= T_s y_{\text{LPF}}^*(-\ell T_s), \end{aligned} \quad (5)$$

where the first equality follows from the fact that $t \mapsto \operatorname{sinc}(t)$ is symmetric (even). Inserting (5) in the RHS of (4), we finally obtain

$$\langle \mathbf{x}, \mathbf{y} \rangle = T_s \sum_{\ell=-\infty}^{\infty} x(-\ell T_s) y_{\text{LPF}}^*(-\ell T_s) = T_s \sum_{\ell=-\infty}^{\infty} x(\ell T_s) y_{\text{LPF}}^*(\ell T_s).$$

Problem 6

Inner Product between Passband Signals

By Theorem 7.7.12,

$$\langle \mathbf{x}_{\text{PB}}, \mathbf{y}_{\text{PB}} \rangle = 2 \operatorname{Re} \left(\langle \mathbf{x}_{\text{BB}}, \mathbf{y}_{\text{BB}} \rangle \right). \quad (6)$$

Also by Theorem 7.7.12, \mathbf{x}_{BB} and \mathbf{y}_{BB} are energy-limited signals that are bandlimited to $W/2$ Hz. Thus, by the Sampling Theorem,

$$\langle \mathbf{x}_{BB}, \mathbf{y}_{BB} \rangle = T \sum_{\ell=-\infty}^{\infty} x_{BB}(\ell T) y_{BB}^*(\ell T), \quad (7)$$

where $T = 1/W$. Substituting (7) into (6) gives the desired equation.