

# Communication and Detection Theory

Signal and Information  
Processing Laboratory

Institut für Signal- und  
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## Model Answers to Exercise 4 of March 14, 2017

<http://www.isi.ee.ethz.ch/teaching/courses/cdt>

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### Problem 1

### *A Specific Signal*

Denote the baseband representation of  $\mathbf{x}$  by  $\mathbf{x}_{\text{BB}}$ . We are given that

$$\mathbf{x}_{\text{BB}}\left(\frac{\ell}{W}\right) = \begin{cases} 1 + i & \text{if } \ell = 0, \\ 0 & \text{otherwise,} \end{cases} \quad \ell \in \mathbb{Z}.$$

Thus, the passband signal  $\mathbf{x}$  is given by

$$\begin{aligned} x(t) &= 2 \sum_{\ell=-\infty}^{\infty} \operatorname{Re}\left(e^{i2\pi f_c t} x_{\text{BB}}\left(\frac{\ell}{W}\right)\right) \operatorname{sinc}(Wt - \ell) \\ &= 2 \operatorname{Re}\left(e^{i2\pi f_c t} (1 + i)\right) \operatorname{sinc}(Wt) \\ &= 2\left(\cos(2\pi f_c t) - \sin(2\pi f_c t)\right) \operatorname{sinc}(Wt), \quad t \in \mathbb{R}. \end{aligned}$$

### Problem 2

### *Multiplying by a Carrier*

Define  $\mathbf{y}: t \mapsto x(t) \cos(2\pi f_c t)$  so

$$\hat{y}(f) = \frac{1}{2} \left( \hat{x}(f - f_c) + \hat{x}(f + f_c) \right), \quad f \in \mathbb{R},$$

from which it follows that

$$\hat{y}_{\text{BB}}(f) = \frac{1}{2} \hat{x}(f), \quad f \in \mathbb{R}$$

and hence

$$y_{\text{BB}}(t) = \frac{1}{2} x(t), \quad t \in \mathbb{R}.$$

Thus, the complex samples of  $\mathbf{y}$  are given by

$$y_{\text{BB}}\left(\frac{\ell}{W}\right) = \frac{1}{2} x\left(\frac{\ell}{W}\right), \quad \ell \in \mathbb{Z}.$$

Next, define  $\mathbf{u}: t \mapsto x(t) \sin(2\pi f_c t)$ , so

$$\hat{u}(f) = -\frac{i}{2} \left( \hat{x}(f - f_c) - \hat{x}(f + f_c) \right), \quad f \in \mathbb{R},$$

from which it follows that

$$\hat{u}_{\text{BB}}(f) = -\frac{i}{2} \hat{x}(f), \quad f \in \mathbb{R},$$

and hence

$$u_{\text{BB}}(t) = -\frac{i}{2} x(t), \quad t \in \mathbb{R}.$$

Thus, the complex samples of  $\mathbf{u}$  are given by

$$u_{\text{BB}}\left(\frac{\ell}{W}\right) = -\frac{i}{2} x\left(\frac{\ell}{W}\right), \quad \ell \in \mathbb{Z}.$$

### Problem 3

### *Orthogonal Passband Signals*

Since the inner product between  $\mathbf{x}_{\text{PB}}$  and  $\mathbf{y}_{\text{PB}}$  is given by

$$\langle \mathbf{x}_{\text{PB}}, \mathbf{y}_{\text{PB}} \rangle = \frac{2}{W} \operatorname{Re} \left( \sum_{\ell=-\infty}^{\infty} x_{\text{BB}}\left(\frac{\ell}{W}\right) y_{\text{BB}}^*\left(\frac{\ell}{W}\right) \right),$$

it follows that the two passband signals  $\mathbf{x}_{\text{PB}}$  and  $\mathbf{y}_{\text{PB}}$  are orthogonal if, and only if,

$$\operatorname{Re} \left( \sum_{\ell=-\infty}^{\infty} x_{\text{BB}}\left(\frac{\ell}{W}\right) y_{\text{BB}}^*\left(\frac{\ell}{W}\right) \right) = 0.$$

### Problem 4

### *The Convolution Revisited*

Let  $\mathbf{z} = \mathbf{x} \star \mathbf{y}$ . By Proposition 7.6.12,  $\mathbf{z}$  is a real integrable passband signal that is bandlimited to  $W$  Hz around the carrier frequency  $f_c$  and whose baseband representation is

$$\mathbf{z}_{\text{BB}} = \mathbf{x}_{\text{BB}} \star \mathbf{y}_{\text{BB}},$$

where  $\mathbf{x}_{\text{BB}}$  and  $\mathbf{y}_{\text{BB}}$  are the baseband representations of  $\mathbf{x}$  and  $\mathbf{y}$ . The complex samples of  $\mathbf{z}$  can now be written as

$$\begin{aligned} \mathbf{z}_{\text{BB}}\left(\frac{\ell}{W}\right) &= \int_{-\infty}^{\infty} x_{\text{BB}}(\tau) y_{\text{BB}}\left(\frac{\ell}{W} - \tau\right) d\tau \\ &= \left\langle \mathbf{x}_{\text{BB}}, t \mapsto y_{\text{BB}}^*\left(\frac{\ell}{W} - t\right) \right\rangle. \end{aligned} \quad (1)$$

Since  $\mathbf{x}_{\text{BB}}$  and  $\mathbf{y}_{\text{BB}}$  are integrable signals bandlimited to  $W/2$  Hz they are also energy-limited (see Note 6.4.12). Moreover, if  $\mathbf{y}_{\text{BB}}$  is energy-limited and bandlimited to  $W/2$  Hz, then so is the signal  $t \mapsto y_{\text{BB}}^*(\ell/W - t)$ . Consequently, we can use Part (iii) of the Sampling Theorem (Theorem 8.4.3) to express the inner product in (1) in terms of the complex samples of  $\mathbf{x}$  and  $\mathbf{y}$ :

$$\left\langle \mathbf{x}_{\text{BB}}, t \mapsto y_{\text{BB}}^*\left(\frac{\ell}{W} - t\right) \right\rangle = \frac{1}{W} \sum_{\ell'=-\infty}^{\infty} x_{\text{BB}}\left(\frac{\ell'}{W}\right) y_{\text{BB}}\left(\frac{\ell}{W} - \frac{\ell'}{W}\right).$$

We conclude that the complex samples of  $\mathbf{z}$  are obtained by convolving the complex samples of  $\mathbf{x}$  with the complex samples of  $\mathbf{y}$  and scaling the result by  $1/W$ :

$$\mathbf{z}_{\text{BB}}\left(\frac{\ell}{W}\right) = \frac{1}{W} \sum_{\ell'=-\infty}^{\infty} x_{\text{BB}}\left(\frac{\ell'}{W}\right) y_{\text{BB}}\left(\frac{\ell - \ell'}{W}\right).$$

**Problem 5*****Exploiting Orthogonality***

The numbers  $X^{(1)}$  and  $X^{(2)}$  can be recovered from  $\mathbf{X}$  by computing linear combinations of the inner products  $\langle \mathbf{X}, \phi_1 \rangle$  and  $\langle \mathbf{X}, \phi_2 \rangle$ . To see this, first note that

$$\langle \mathbf{X}, \phi_1 \rangle = \left( A^{(1)} X^{(1)} + A^{(2)} X^{(2)} \right) \|\phi_1\|_2^2,$$

and that

$$\langle \mathbf{X}, \phi_2 \rangle = \left( A^{(1)} X^{(1)} - A^{(2)} X^{(2)} \right) \|\phi_2\|_2^2,$$

where in the computation of the inner products we have used that  $\langle \phi_1, \phi_2 \rangle = 0$ . It thus follows that

$$\begin{aligned} X^{(1)} &= \frac{1}{2A^{(1)}} \left( \frac{\langle \mathbf{X}, \phi_1 \rangle}{\|\phi_1\|_2^2} + \frac{\langle \mathbf{X}, \phi_2 \rangle}{\|\phi_2\|_2^2} \right) \\ X^{(2)} &= \frac{1}{2A^{(2)}} \left( \frac{\langle \mathbf{X}, \phi_1 \rangle}{\|\phi_1\|_2^2} - \frac{\langle \mathbf{X}, \phi_2 \rangle}{\|\phi_2\|_2^2} \right). \end{aligned}$$