

Communication and Detection Theory

Signal and Information
Processing Laboratory

Institut für Signal- und
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Model Answers to Exercise 6 of March 28, 2017

<http://www.isi.ee.ethz.ch/teaching/courses/cdt>

Problem 1

Superimposing Independent Transmissions

Since the data streams $(D_j^{(1)})$ and $(D_j^{(2)})$ are independent, it follows that the sequences of symbols $(X_\ell^{(1)})$ and $(X_\ell^{(2)})$ are independent. Consequently, the stochastic processes $(X^{(1)}(t))$ and $(X^{(2)}(t))$ are independent. Therefore, *a fortiori*, at every epoch $t \in \mathbb{R}$ the random variables $X^{(1)}(t)$ and $X^{(2)}(t)$ are independent and hence uncorrelated. Since they are both of zero mean,

$$\mathbb{E}[X^{(1)}(t) X^{(2)}(t)] = 0, \quad t \in \mathbb{R}.$$

The power P in the sum of the processes is thus

$$\begin{aligned} P &= \lim_{T \rightarrow \infty} \frac{1}{2T} \mathbb{E} \left[\int_{-T}^T (X^{(1)}(t) + X^{(2)}(t))^2 dt \right] \\ &= \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T \mathbb{E} \left[(X^{(1)}(t) + X^{(2)}(t))^2 \right] dt \\ &= \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T \left(\mathbb{E} \left[(X^{(1)}(t))^2 \right] + \mathbb{E} \left[(X^{(2)}(t))^2 \right] \right) dt \\ &= \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T \mathbb{E} \left[(X^{(1)}(t))^2 \right] dt + \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T \mathbb{E} \left[(X^{(2)}(t))^2 \right] dt \\ &= p^{(1)} + p^{(2)}. \end{aligned}$$

Problem 2

The Minimum Distance of a Constellation and Power

Using Theorem 14.5.2 we obtain

$$\begin{aligned} P &= \frac{A^2}{T_s} \lim_{L \rightarrow \infty} \frac{1}{2L+1} \sum_{\ell=-L}^L \mathbb{E}[X_\ell^2] \\ &= \frac{A^2}{T_s} \mathbb{E}[X_0^2], \end{aligned}$$

where the second equality follows from the hypothesis that the symbols (X_ℓ) are IID and *a fortiori* of the same second moment. It thus remains to compute this second moment:

$$\begin{aligned} \mathbb{E}[X_0^2] &= \frac{1}{\nu} \frac{d^2}{4} \left(1^2 + 3^2 + \dots + (2\nu - 1)^2 \right) \\ &= \frac{1}{\nu} \frac{d^2}{4} \frac{(2\nu + 1)(2\nu - 1)\nu}{3} \\ &= \frac{1}{12} d^2 (2\nu + 1)(2\nu - 1). \end{aligned}$$

Hence,

$$P = \frac{A^2}{T_s} d^2 \frac{(2\nu + 1)(2\nu - 1)}{12}.$$

Problem 3

PAM with Nonorthogonal Pulses

Using (14.37) and noting that for the repetition code $X_1 = X_2$, we obtain

$$\begin{aligned} P &= \frac{1}{2T_s} \mathbb{E} \left[\int_{-\infty}^{\infty} \left(A \sum_{\ell=1}^2 X_\ell g(t - \ell T_s) \right)^2 dt \right] \\ &= \frac{1}{2T_s} \mathbb{E} \left[\int_{-\infty}^{\infty} A^2 X_1^2 \left(g(t - T_s) + g(t - 2T_s) \right)^2 dt \right] \\ &= \frac{1}{2T_s} \mathbb{E} \left[\int_{-\infty}^{\infty} A^2 X_1^2 \left(g(t) + g(t - T_s) \right)^2 dt \right] \\ &= \frac{1}{2T_s} A^2 \mathbb{E}[X_1^2] \int_{-\infty}^{\infty} \left(\mathbb{I}\{|t| \leq T_s\} + \mathbb{I}\{|t - T_s| \leq T_s\} \right)^2 dt \\ &= \frac{A^2}{2T_s} \int_{-\infty}^{\infty} \left(\mathbb{I}\{|t| \leq T_s\} + \mathbb{I}\{|t - T_s| \leq T_s\} \right)^2 dt \\ &= \frac{A^2}{2T_s} 6T_s \\ &= 3A^2. \end{aligned}$$

Problem 4

Non-IID Data Bits

The following example demonstrates that if the data bits are not IID then (14.37) need not hold. Suppose that the symbols (X_ℓ) are produced from the data bits (D_j) according to the antipodal rule

$$0 \mapsto 1, \quad 1 \mapsto -1.$$

Further assume that the data bits are not IID but that instead the odd data bits are IID random bits but $D_{2\ell} = D_{2\ell-1}$ for all $\ell \in \mathbb{Z}$. Thus,

$$\dots, D_3, D_{-1}, D_1, D_3, D_5, \dots,$$

are IID random bits but

$$\dots, D_0 = D_{-1}, \quad D_2 = D_1, \quad D_4 = D_3, \dots$$

Further assume the pulse shape $\mathbf{g}: t \mapsto \mathbb{I}\{|t| \leq T_s\}$, where T_s is the baud period. This scenario can be viewed as though the IID random bits (D_{2j+1}) were fed to a blocklength $N = 2$ repetition code as in Exercise 14.3. From that exercise we conclude that the transmitted power is

$$P = 3A^2.$$

However, if we ignored the dependence and computed the RHS of (14.37) (with $N = 1$) we would get $2A^2$, which is clearly not the transmitted power.

Problem 5

The Power in Nonorthogonal PAM

(i) By (14.33) the power in this case is given by

$$\begin{aligned} P &= \frac{1}{T_s} A^2 \|\mathbf{g}\|_2^2 \\ &= \frac{1}{T_s} A^2 2T_s \\ &= 2A^2. \end{aligned}$$

(ii) Using (14.31) the power in this case is given by

$$\begin{aligned} P &= \frac{1}{T_s} A^2 \left(K_{XX}(-1) R_{\mathbf{g}\mathbf{g}}(-T_s) + K_{XX}(0) R_{\mathbf{g}\mathbf{g}}(0) + K_{XX}(1) R_{\mathbf{g}\mathbf{g}}(T_s) \right) \\ &= \frac{1}{T_s} A^2 \left(\frac{1}{2} T_s + 1(2T_s) + \frac{1}{2} T_s \right) \\ &= 3A^2. \end{aligned}$$

Problem 6

Pre-Encoding

This does not change the transmitted power. Indeed, since the original data bits D_1, \dots, D_K are IID random bits, it follows that the K -tuple (D_1, \dots, D_K) is uniformly distributed over the set $\{0, 1\}^K$ of all binary K -tuples. Since ϕ is a one-to-one mapping, it follows that the K -tuple (D'_1, \dots, D'_K) is also uniformly distributed over $\{0, 1\}^K$, so D'_1, \dots, D'_K are also IID random bits. Thus, the statistics of the bits fed to the encoder in the two cases are the same, and the power is thus the same.